

Sample intervals

Design and performance of fast transient detectors

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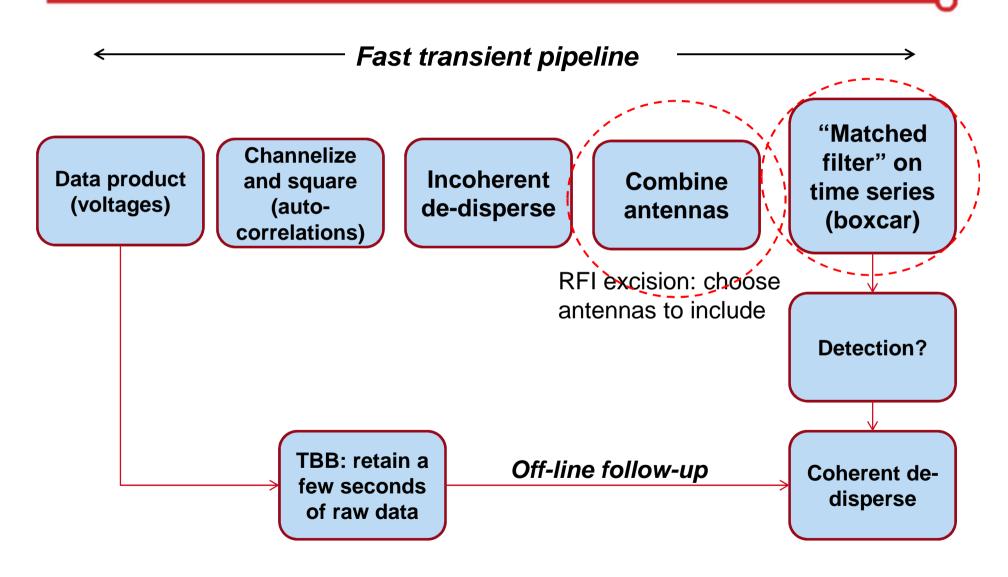


Outline

- ➤ The incoherent detector
 - The classical matched filter pros and cons
 - Degrading effects scattering, inaccurate de-dispersion, trial templates
 - Performance for high DM events, at different frequencies
- ➤ How to design a better detector?
 - CRAFT detector working with the dynamic spectrum
 - Fitting into the hierarchy and implementation
 - Asymptotic performance



Fast transient source detection



The Matched Filter

- Optimal detector for a *known* signal in *known* Gaussian noise
- Matches expected signal shape (template, h) to data (s), and sums

$$T(x) = (s \star h)[n] = \sum_{k=0}^{K} s[k]h[n-k],$$

- Pros: optimal for a given Gaussian dataset
- Cons: requires full signal knowledge, not "blind" to signal shape
- Performance: signal-to-noise ratios

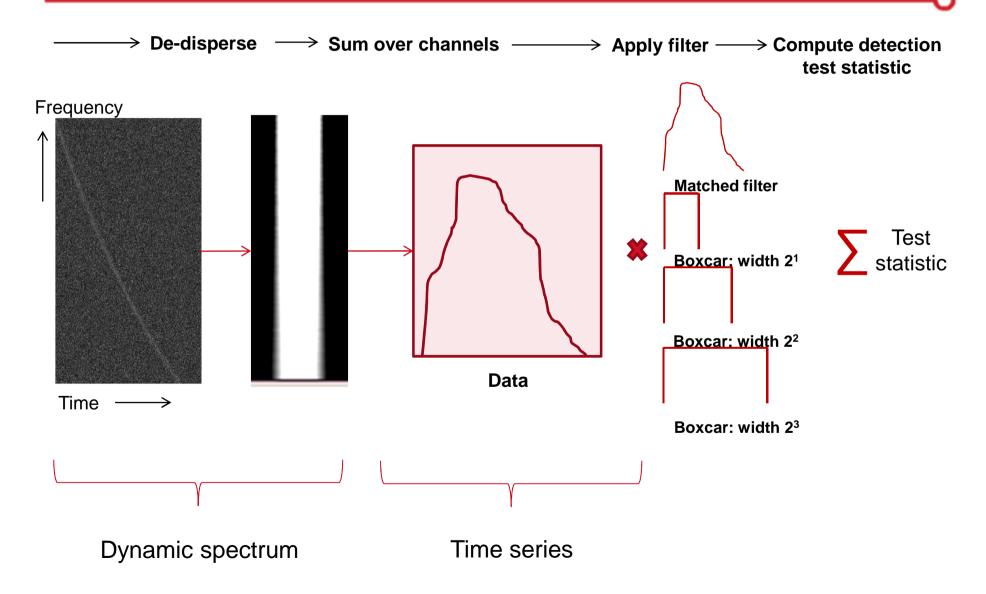
$$d = \frac{\sqrt{\int_W s(t)^2 dt}}{\sigma}. \quad \text{Template (h)} \quad d = \frac{\int_W s(t)h(t)dt}{\sigma\sqrt{\int_W h(t)^2 dt}},$$

$$\min_{s, matched} s = \frac{\int_W s(t)h(t)dt}{\sigma\sqrt{\int_W h(t)^2 dt}},$$

$$d = \frac{\int_W s(t)h(t)dt}{\sigma\sqrt{\int_W h(t)^2 dt}}$$



Time series MF implementation





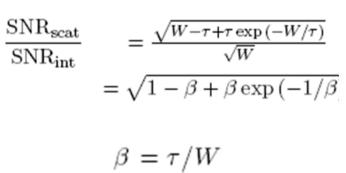
Degrading factors

- Intrinsic
 - Scatter broadening due to ISM multi-path
- Extrinsic
 - Incorrect dispersion measure
 - Finite temporal/spectral sampling
 - Finite temporal window
 - Mis-matched/approximate templates

Degrading factors: ISM scattering

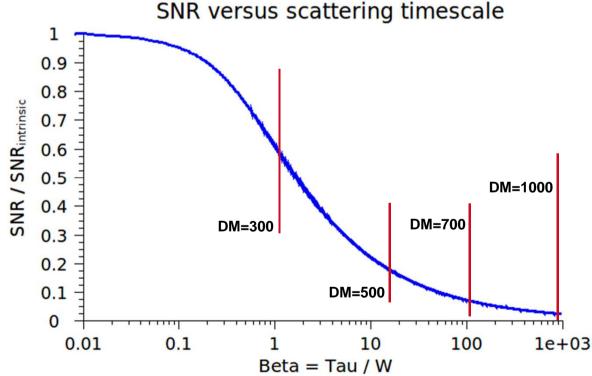
Characteristic scattering timescale:

$$\log \tau_d = -3.72 + 0.411 \log \mathrm{DM} + 0.937 (\log \mathrm{DM})^2 - 4.4 \log \nu_{\mathrm{GHz}} \ \mu \mathrm{s}.$$
 Cordes & Lazio (2002)



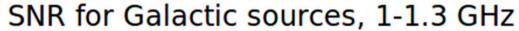
ASKAP parameters: W = 1 ms

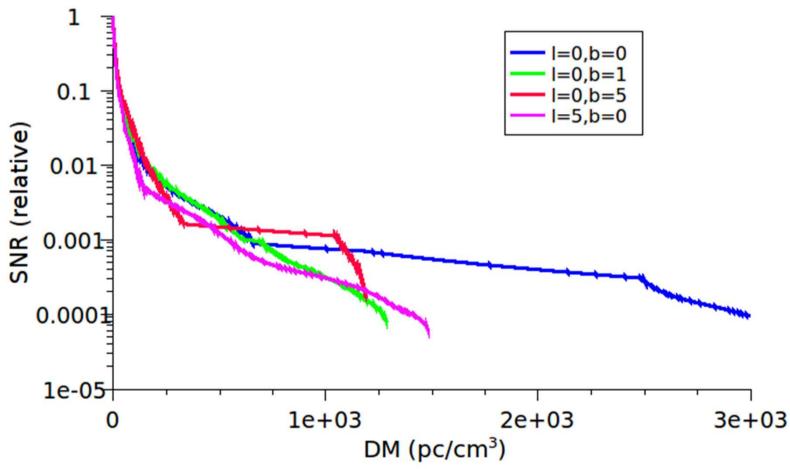
v = 1 GHz



Impact of degrading factors: Galactic lines-of-sight

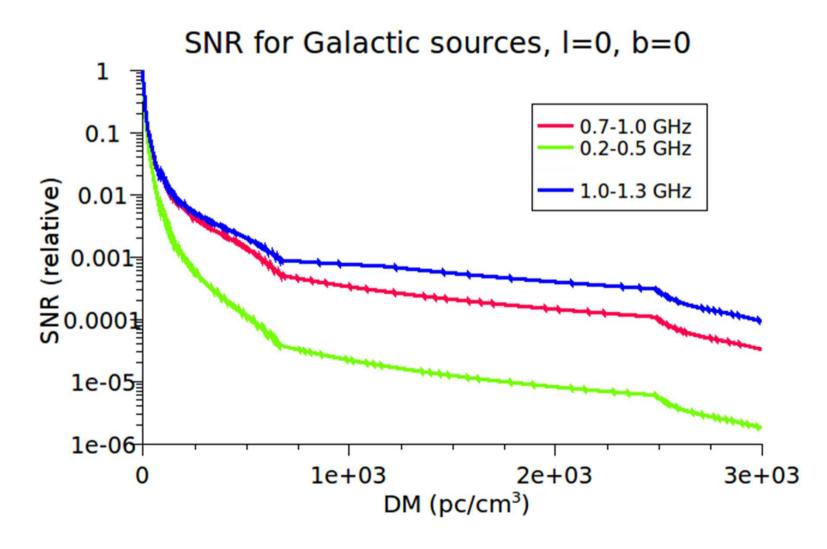
Relative detection performance for identical source at different DMs:





Impact of degrading factors: Frequency dependence

Relative detection performance for identical source at different DMs:



Hierarchy of signal knowledge

Dynamic spectrum power samples

Matched filter

Full signal knowledge required: temporal and spectral

$$SNR_{DSMF} = \frac{\sqrt{\sum_{W} \sum_{i=1}^{N_t} \bar{P}^2(t, \nu_i)} \sqrt{\Delta t}}{kT_{sys} \sqrt{\Delta \nu}}.$$

Optimized boxcar templates

Partially blind: trial pulse widths, potentially trial spectral index

$$\mathrm{SNR}_{\mathrm{CM}} = \frac{P_{\mathrm{dedisp}}}{P_{N_{\mathrm{dedisp}}}} = \sqrt{\frac{\Delta t}{N_{\mathbb{S}} \Delta \nu}} \frac{\sum_{s \in \mathbb{S}} \bar{P}(t_s, \nu_s)}{k T_{\mathrm{sys}}}.$$

Time series power (summed over spectral channels)

Matched filter

Full time domain signal knowledge required

$$SNR_{MF} = \frac{\sqrt{\sum_{W} \left(\sum_{i=1}^{N_t} \bar{P}(t, \nu_i)\right)^2} \sqrt{\Delta t}}{kT_{sys}\sqrt{\Delta \nu}\sqrt{N_{ch}}}.$$

Boxcar templates

Blind: coarse trial of pulse widths

$$\mathrm{SNR}_{\mathrm{BMF}} = \sqrt{\frac{\Delta t}{N_{\mathrm{ch}} W \Delta \nu}} \frac{\sum\limits_{i=1}^{N_{\mathrm{t}}} \bar{P}(t, \nu_{i})}{k T_{\mathrm{sys}}}.$$

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Boxcar templates

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Hierarchy of signal knowledge

Dynamic spectrum power samples

Matched filter

Full signal knowledge required: temporal and spectral

$$SNR_{DSMF} = \frac{\sqrt{\sum_{W} \sum_{i=1}^{N_t} \bar{P}^2(t, \nu_i)} \sqrt{\Delta t}}{kT_{sys} \sqrt{\Delta \nu}}.$$

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Boxcar templates

Blind: coarse trial of pulse widths

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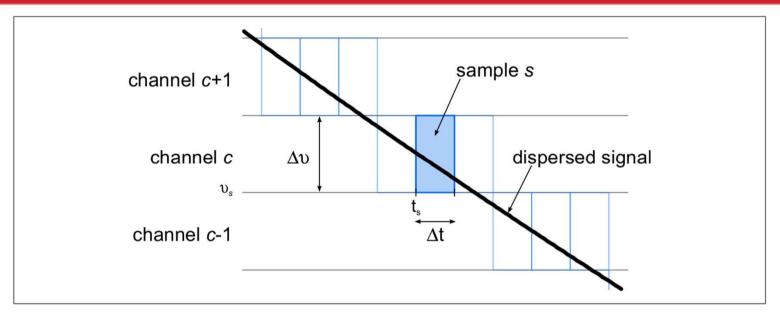
Dynamic spectrum detection

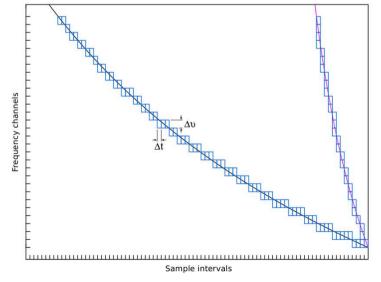
- Design of fast transient detector for CRAFT, using FPGAs to implement de-dispersion and detection algorithm
- Works directly with dynamic spectrum power samples from spectrometer
- Detector:
 - Choose samples according to expected power for a given DM: set spectral index (0), pulse width
 - Average power in a sample calculated analytically
 - Choose set of samples that maximises signal-to-noise ratio
 - "Sample-optimised" boxcar template

Clarke et al. (2011, in prep)



Dynamic spectrum detection





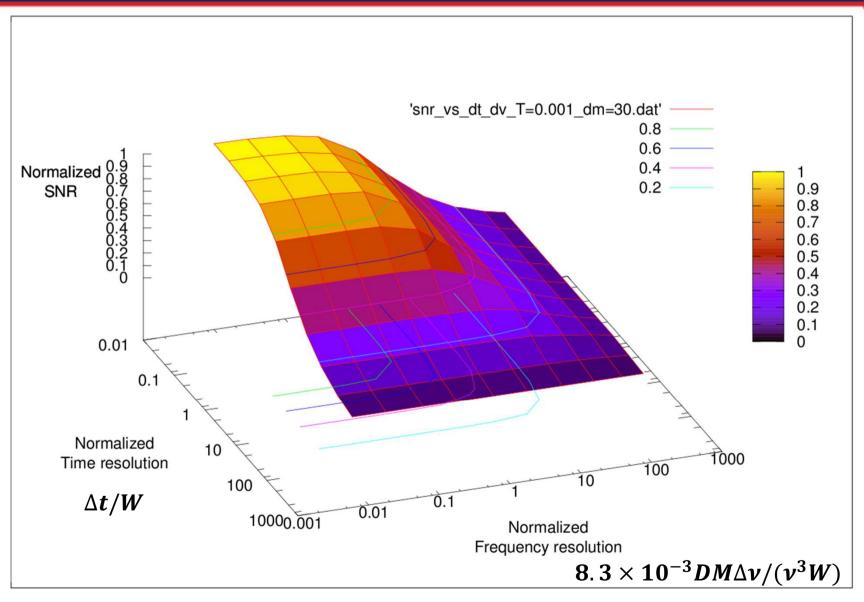
Inclusion criterion:

$$\bar{P}(t_{\varsigma}, \nu_{\varsigma}) > \left(\sqrt{\frac{N_{\mathbb{S}} + 1}{N_{\mathbb{S}}}} - 1\right) \sum_{s \in \mathbb{S}, s \neq \varsigma} \bar{P}(t_{s}, \nu_{s})$$

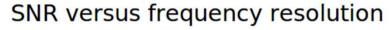
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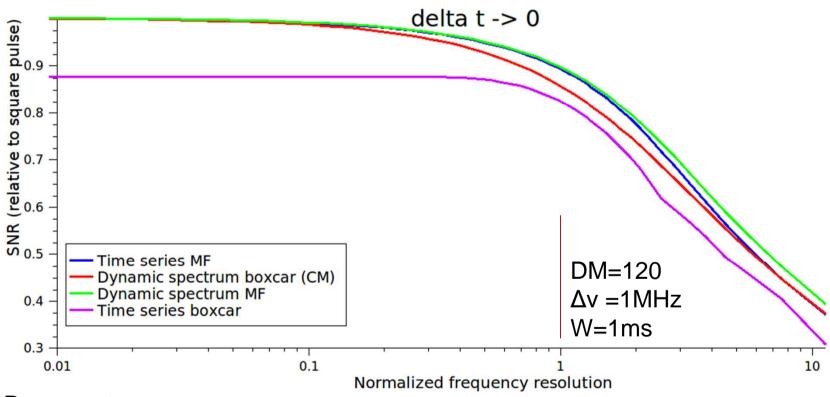


Asymptotic performance



Performance comparison





ASKAP parameters:

$$v = [1.0,1.3] \text{ GHz}$$

 $\Delta v_{TOT} = 300 \text{ MHz}$

$$8.3\times10^{-3}DM\Delta\nu/(\nu^3W)$$



Summary

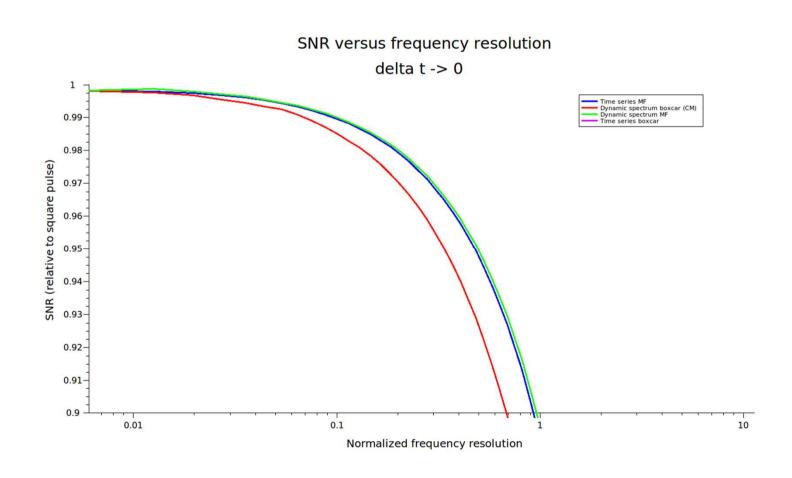
- High DM detections difficult, particularly at low frequencies
- Optimal matched filter rarely achievable ----- smart boxcar template applied to dynamic spectrum dataset can recover some lost performance

Next steps:

- › Balance combination of DM steps / spectral index steps / pulse width steps for optimal detection with a given FPGA design and architecture
- Compare performance with other FT experiments (e.g., VFASTR)



Performance comparison





Performance comparison

